

Periodic Operation of Linear Circuits (Non-Sinusoidal)

A linear circuit operates under periodic state conditions if all electromagnetic quantities associated with its operation present a periodic time-variation of the same frequency or, equivalently, with the same period T .

$$f(t) = f(t + nT)$$

The study of the periodic operation of linear circuits can be done by invoking the superposition theorem: periodic time-varying functions are approached as combinations appropriately chosen simpler periodic functions of time. And the immediate candidates simpler periodic functions are the harmonic time-varying functions!

$$\omega = \frac{2\pi}{T} \rightarrow \text{fundamental frequency}$$

It can be proved that a periodic time-varying function, which is square-integrable (quadratically) integrable and sectionally smooth over a period can be expressed as an linear combination of the above specified basis functions. An equivalent statement is valid under somewhat different conditions, namely for periodic time-varying functions which are absolutely integrable and sectionally smooth over a period.

A periodic time-varying function of period T which satisfies the above enumerated conditions admits the Fourier series expansion.

$$f(t) = \frac{1}{2}(a_0 + b_0) + \sum_{k=1}^{\infty} (a_k \cos k\omega t + b_k \sin k\omega t)$$

$$a_k = \frac{2}{T} \int_{\theta}^{\theta+T} f(t) \cos k\omega t \, dt \quad b_k = \frac{2}{T} \int_{\theta}^{\theta+T} f(t) \sin k\omega t \, dt \quad k \in \mathbb{N}$$

$$\frac{a_0}{2} = \frac{1}{T} \int_{\theta}^{\theta+T} f(t) \cos 0 \, dt = \frac{1}{T} \int_{\theta}^{\theta+T} f(t) \, dt$$

$$\frac{b_0}{2} = \frac{1}{T} \int_{\theta}^{\theta+T} f(t) \sin 0 \, dt = 0$$

$$f(t) = F_0 + \sum_{k=1}^{\infty} F_k \sqrt{2} \sin(k\omega t + \psi_k) = \sum_{k=0}^{\infty} f_k(t)$$

$$F_0 = \frac{a_0}{2} = \frac{1}{T} \int_{\theta}^{\theta+T} f(t) \, dt$$

$$F_k = \sqrt{\frac{a_k^2 + b_k^2}{2}}$$

In particular, the average value $f_0 = F_0$ might be named the harmonic of order zero

The harmonic components are linearly independent

The set $\{f_k\}_{k=0,1,2}$ of harmonic components associated with fund. frequency ω are orthogonal to each other with respect to the so-called scalar product

$$\langle f, g \rangle = \frac{1}{T} \int_{\theta}^{\theta+T} f g \, dt = \overline{fg}$$

Harmonic analysis of the periodic equation

The periodic equation of a linear el. circuit is described by the set of Kirchhoff's equations.

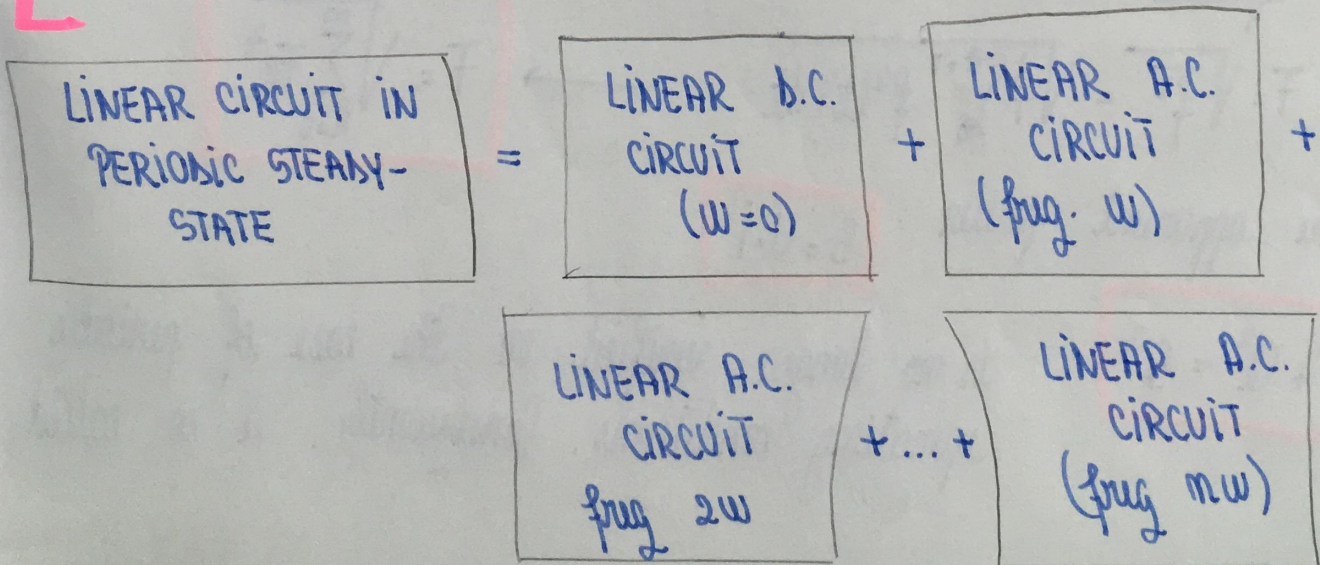
$$\left\{ \begin{aligned} i_h(t) &= \sum_{m=1}^{\infty} i_h^m(t) & ; & & u_{th}(t) &= \sum_{m=1}^{\infty} u_{th}^m(t) \\ \sum_{h \in (a)} i_h(t) &= 0 \quad a=1, N-1 & ; & & \sum_{h \in (p)} u_{th}(t) &= 0 \quad p=1, B-N+1 \end{aligned} \right.$$

Kirchhoff's equations have to be satisfied separately for each harmonic component.

Kirchhoff's voltage theorem is to be completed with Joule's theorem, for each harmonic component:

$$m=0 \quad U_{tk}^0 = \begin{cases} -U_k^0 - E_k^0 & \text{for a branch including a capacitor} \\ R_k i_k^0 - U_k^0 - E_k^0 & \text{for a branch without a capacitor} \end{cases}$$

$$U_{th}^m = \sum_{h=1}^L Z_{hs}^m I_h^m - U_k^m - E_k^m, \quad m=1, 2, \dots$$



The algorithm for calculation is:

- 1°. The harmonic Fourier series expansions are computed
- 2°. The circuit is studied separately for the direct current component $m=0$, and for successive harmonic components $m=1, 2, \dots$ to find the harmonic components of the unknown currents and voltages.
- 3°. Each unknown current & voltage is synthesised as the sum of its computed harmonic components.

The measured active power: $P = \overline{u \cdot i} = \frac{1}{T} \int_{\theta}^{\theta+T} u(t) i(t) dt$

$$P = \overline{u \cdot i} = \left(U_0 + \sum_{k=1}^{\infty} u_k \right) \left(I_0 + \sum_{k=1}^{\infty} i_k \right)$$

$$P = U_0 I_0 + \sum_{k=1}^{\infty} U_k I_k \cos \psi_k$$

Reactive power

$$Q = \sum_{k=1}^{\infty} U_k I_k \sin \psi_k$$

The effective value:

$$F = \sqrt{\overline{f^2}} = \sqrt{\frac{1}{T} \int_{\theta}^{\theta+T} f^2(t) dt}$$

$$F = \sqrt{\sum_{k=0}^{\infty} F_k^2}$$

The apparent power:

$$S = U \cdot I$$

$$P^2 + Q^2 = S^2$$

is no longer verified in the case of periodic operating conditions. Consequently, a so called

distortion power can be defined as:

$$D = \sqrt{S^2 - (P^2 + Q^2)}$$

The distortion residual

$$F_d = \sqrt{\sum_{k=1}^{\infty} F_k^2} = \sqrt{F^2 - F_0^2 - F_1^2}$$

$$F_d = \sqrt{\sum_{k=1}^{\infty} F_k^2} = \sqrt{F^2 - F_1^2}$$

The distortion coefficient:

$$hd = \frac{F_d}{F}$$